Design of a Relocatable Antenna Element Module for a Ku-Band Reconfigurable Array with Low Mutual Coupling

Seongjung Kim · Sangwook Nam*

Abstract

In this study, a modular array antenna with low mutual coupling is proposed for operation in a Ku-band (12–18 GHz) reconfigurable array, which consists of a ground frame board, element blocks, spacing blocks, and connecting rods. The modular element blocks can be freely mounted at any positions on a 46×9 grid on the ground frame board. The proposed modular array structure can be used to easily check the performance of different array structures by changing the element positions due to its low mutual coupling characteristics. An expanded view of the proposed array antenna structure is also provided, showcasing how the antenna modules, ground modules, base ground, ground frames, and aluminum rods have been constructed. The value of mutual coupling measured between the adjacent elements is less than -22.9 dB in the E-plane and -18.5 dB in the H-plane. Moreover, the measured radiation efficiency and gain patterns of the active element are shown.

Key Words: Low Mutual Coupling, Modular Array, Reconfigurable Array, Wideband Array Antenna.

I. INTRODUCTION

Broadband array antennas with low mutual coupling have attracted a lot of attention due to their various industrial applications. However, there are two main challenges in the implementation of these array antennas. First, it is difficult to achieve a reduction in mutual coupling between the antenna elements over a wide bandwidth. Second, in many cases, the antenna arrays operating over a wide impedance bandwidth have threedimensional structures that are difficult to implement [1]. One of the critical problems in the implementation is to obtain a reliable electrical contact between the feed lines and the SMA connector by soldering. In fact, the problem is aggravated in large arrays designed to operate at high frequencies, such as the Ku-band. Additionally, in large arrays, there is a considerable risk of rejection if, after the fabrication, some of the antenna elements are found to be defective. Therefore, to overcome these challenges, the implementation of a modular array antenna that can easily replace the faulty antenna elements is required, especially for large arrays.

Owing to their reconfigurability, the modular array antennas have an additional advantage for the development of array structures for microwave radar systems. In the case of developing an imaging radar, there is no definite solution to determine the optimal positions of the TX and RX antenna elements for the acquisition of the best scattering data from the target due to the

Manuscript received December 21, 2021 ; Revised March 2, 2022 ; Accepted April 5, 2021. (ID No. 20211221-156J) School of Electrical and Computer Engineering, Seoul National University, Seoul, Korea. *Corresponding Author: Sangwook Nam (e-mail: snam@snu.ac.kr)

This is an Open-Access article distributed under the terms of the Creative Commons Attribution Non-Commercial License (http://creativecommons.org/licenses/by-nc/4.0) which permits unrestricted non-commercial use, distribution, and reproduction in any medium, provided the original work is properly cited.

 $[\]odot\,$ Copyright The Korean Institute of Electromagnetic Engineering and Science.

dynamic range and direction of the target [2]. Moreover, the reconfigurable array structures can be experimentally used to control the radiation patterns of the arrays and reduce the number of antenna elements in an array using the array thinning technique [3]. In previous research, a wideband modular array was studied, but it had high mutual coupling [4]. In this study, a wideband and low mutually coupled modular array antenna is proposed. The individual antenna element of the proposed array is designed using the methodology reported in [1].

II. DESIGN OF ARRAY ANTENNA STRUCTURE AND OPERATION

To implement the wideband array antenna, grounded electrical side walls must be tightly coupled to the tips of the dipole arms, as shown in Fig. 1(a), and the frequency-selective surface (FSS) in Fig. 1(b) should have proper patch sizes f, g, and h (Table 1). The FSS replaces the dielectric superstrate in [1] and is more suitable for modular unit cells due to its light weight and com-



Fig. 1. Proposed unit cell structure. (a) Perspective view. (b) Front side and (c) back side of the proposed PCB. (d) Side wall with resistive sheets. All dimensions are presented in Table 1.

Table 1. Dimensions of the proposed unit cell (unit: mm)

Parameter	Value	Parameter	Value
а	20	n	0.737
Ь	10	0	1.5
С	10.7	P	1.5
d	21	q	2
е	7.9	r	0.737
f	1.836	S	2
g	0.5	t	3
Н	2	T	6
Ι	2.43	U	2.5
J	3.5	υ	0.5
k	2.1	W	300
l	0.2	у	500
т	0.4	z	3.27

pactness but has the same electromagnetic effect as the dielectric superstrate. Furthermore, the positions and sizes of the resistive sheets determine the quantity of mutual coupling between the antenna elements and the radiation efficiency. The determination of the positions and sizes and a parametric study are described in detail in [1]. The modified unit cell structure consists of side walls, a printed circuit board (PCB), and an aluminum ground, which can be assembled or detached. Fig. 1(a) shows the assembled structure of the proposed antenna element. It features a Tshaped aluminum ground, and the vertical part of this ground structure provides a mechanical support for the PCB from the back. The horizontal part of the ground is slotted in the center to insert the PCB. Two holes are used to insert the aluminum rods, as shown in Fig. 2(b), to fix the unit cell structures. On the bottom side of the PCB, there are two holes for using the endlaunch connector (https://www.with-wave.com/end-launchnarrow-block). The PCB is a Taconic TLY-5 ($\varepsilon_r = 2.2$, tan $\delta =$ 0.0009) with a thickness of 0.25 mm. The simulated resistive film was 250 Ω /square (https://ohmega.com/) and it as laminated on the side wall with FR-4 ($\varepsilon_r = 4.3$, tan $\delta = 0.025$).

Fig. 2(a) shows the proposed 47 antenna elements assembled on a 46×9 ground board. The ground modules, which have two different sizes, fill the positions where there are no antenna modules present. The antenna modules, which are arrays designed using the proposed single element shown in Fig. 1(a), are inserted through the slots of the base ground. As shown in Fig. 2(b), the aluminum rods fix the entire structure by passing through the side holes of the ground frames, antenna modules, and ground modules, and this is the final step of the assembly. The right ends of the rods and the right side of the ground frame are bolted together. The fabricated array antenna is shown in Fig. 3. In this study, only the performances of six ports are presented as a representative, as shown in Fig. 3(a).

Fig. 4 shows the simulated and measured S-parameters of the



Fig. 2. Proposed modular array antenna. Antenna elements are arranged on a one-dimensional array and some elements are sparsely placed around it. (a) Perspective view. (b) Expended view. Antenna element spacing is b.



Fig. 3. Fabricated proposed modular array antenna. (a) Front view of the full array structure. Six port numbers are indicated. (b) Front view of the antenna element. (c) Bottom side of the antenna element, which is the connector line.



Fig. 4. Simulated (solid) and measured (dashed) S-parameters of the six ports in Fig. 3(a). Simulated S_{11} and S_{22} perfectly overlap each other. Γ_1 (solid-pink) is a simulated active reflection coefficient at port1.

six ports. For ports 5 and 6, the simulated and measured mutual coupling was slightly larger than -20 dB at 12 GHz, and the others were lower. Thus, the active reflection coefficient at port1 is similar to S_{11} . Moreover, the simulated and measured reflection coefficients are below -10 dB at the Ku-band. The simulated and measured active element gains at 12, 15, and 18 GHz are shown in Fig. 5. The beamwidth in the E-plane was approximately 120° and 160° in the H-plane. The radiation efficiency and broadside gain are presented in Fig. 6. The simulated efficiency is greater than 65%, and the measured efficiency is greater than 52%. The simulated gain is greater than 2.2 dBi, and the measured gain is larger than 1 dBi. The difference between the simulation and the measurement is considered to be due to the fabrication error: the height *s* of the *fabricated* optimized resis-



Fig. 5. Simulated and measured active element gain patterns, where the E-plane is the xz-plane and the H-plane is the yz-plane in Fig. 1(a).



Fig. 6. Simulated and measured (a) radiation efficiency and (b) broadside gain.

tive film for low mutual coupling is 3 mm, and not 2 mm.

III. CONCLUSION

In this paper, a modular array antenna was proposed with a low mutual coupling in the entire Ku-band. It was proposed that, by dividing the entire structure into modules, a simple mechanical assembly is enough to fabricate any antenna array structure without the need for soldering. The paper also offered a reliable fabrication method for implementing large array antennas. With the proposed structure, it is possible to replace the faulty antenna elements in the large arrays and to experimentally test various array designs without the requirement for the repeated fabrication of different array patterns. This is because the antenna modules can be put into desired locations on the ground frame.

This work was supported by an Institute of Information & Communications Technology Planning & Evaluation (IITP) grant funded by the Government of Korea (MSIT) and partially supported by the BK21 FOUR program in the Education and Research Program for Future ICT Pioneers at Seoul National University in 2021.

REFERENCES

- S. Kim and S. Nam, "A compact and wideband linear array antenna with low mutual coupling," *IEEE Transactions on Antennas and Propagation*, vol. 67, no. 8, pp. 5695-5699, 2019.
- [2] D. Jang, J. Hur, H. Shim, J. Park, C. Cho, and H. Choo, "Array antenna design for passive coherent location systems with non-uniform array configurations," *Journal of Electromagnetic Engineering and Science*, vol. 20, no. 3, pp. 176-182, 2020.
- [3] S. H. Jung, K. I. Lee, H. S. Oh, H. K. Jung, H. Yang, and Y. S. Chung, "Optimal design of thinned array using a hybrid genetic algorithm," *Journal of Electromagnetic Engineering and Science*, vol. 21, no. 4, pp. 261-269, 2021.
- [4] H. Holter, "Dual-polarized broadband array antenna with BOR-elements, mechanical design and measurements," *IEEE Transactions on Antennas and Propagation*, vol. 55, no. 2, pp. 305-312, 2007.